

A Self-Packaged SISL Low-Power Rectifier Based on a High-Impedance Line for C-Band Applications

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Abstract—In this letter, a self-packaged substrate-integrated suspended line (SISL) rectifier is proposed. The rectifier consists of a quarter-wavelength high-impedance transmission line, a diode, a tuning inductor, and a dc pass filter. The lower the power is, the lower the rectifying efficiency is. At a low input power level, the diode impedance real part is almost zero. A tuning inductor is introduced to form a parallel resonator to enhance the diode voltage. A quarter-wavelength high-impedance line is presented to match the rectifying diode to the source. In addition, due to the self-packaging characteristics and air cavity of SISLs, the radiation loss and dielectric loss of the proposed rectifier are reduced. A rectifier was designed, fabricated, and measured at 5.8 GHz. The radio frequency (RF)-dc conversion efficiency reaches 3.5% at −30 dBm, which is close to its theoretical estimated efficiency at 5.8 GHz. This ultralow power rectifier may be used in energy harvesting at C-band.

Index Terms—High impedance line, rectifier, radio frequency (RF)-dc conversion efficiency, substrate integrated suspended line (SISL), ultralow power.

I. INTRODUCTION

WITH the rapid development of the Internet of Things (IoT), wireless sensor networks, biomedical implant systems, and intelligent transportation systems in recent years, a large number of low-power and miniaturized wireless sensors are distributed in all corners of the world [1], [2]. Using batteries to power sensors increases the size of devices and is not sustainable and practically impossible. Due to the wide application of power sources such as cellular mobile base stations and radio frequency (RF) signal tower, the space is flooded with a large amount of RF energy [3]. Ambient RF energy harvesting is expected to solve the power supply problem of electronic systems in some emerging application fields [4], [5], [6], [7], [8], [9], [10], [11], [12], [13], [14], [15], [16], [17], [18], [19]. Because the power spectrum of signals in the environment is relatively low with a typical power density of $2 \mu\text{W}/\text{m}^2$ – $10 \text{ mW}/\text{m}^2$ [20], the rectifiers are required to have excellent RF-dc conversion performance at a much low input power range.

In addition, in the 1950s, Glaser [21] put forward the solar satellite plan. The solar satellite project hopes to use

TABLE I
PERFORMANCE WITH THE PRIOR RECTIFIERS

Ref.	f (GHz)	Diode	Efficiency @ −40 dBm and −30 dBm			
			−40 dBm		−30 dBm	
			Measure	Estimate	Measure	Estimate
[26]	2.35	Tunnel diode	3.8%	2.1%	18.2%	21.5%
[29]	0.9	SMS7630	3%	4%	17.5%	40.4%
[34]	1.9	VDI ZBD	1.8%	2.7%	11.2%	27.1%
This work	5.8	SMS7630	1.2%	0.3%	3.5%	3.14%

solar energy to power electronic components on the Earth through microwave power transmission, in an effort to solve the resource shortage facing humanity [22]. Low-orbit near-Earth satellites for the space solar power satellite (SSPS) verification are planned to launch in the future. The microwave power transmission may be at 5.8 GHz. The power density on the ground is very low due to the limitation of the transmitting antenna aperture and the transmitted power level. It requires microwave rectifiers working at ultralow power level, e.g., about −30 dBm.

Multifarious effective methods have been adopted to enhance the efficiency of ultralow power rectifiers. For example, hybrid energy collection exacerbating a nonlinear process in the diode was a significant method to increase the RF-dc conversion efficiency of energy harvesters [23], [24]. Also, some diodes with a strong nonlinearity such as tunnel diodes [25], nanodiodes [26], and spin diodes [27] were used to elevate efficiency. However, it is very challenging to break the rectifying efficiency limitation by ameliorating the diode manufacturing process. Moreover, the RF-dc conversion efficiency of rectifiers was raised by making the diodes operate at the optimum temperature [28], [29]. The above rectifiers are all operating in the S-band and below. There are few research studies on C-band with a typical power of −30 dBm and below, where conversion efficiency is so limited.

The substrate-integrated suspended line (SISL) technology can reduce radiation loss and dielectric loss significantly [30], and has been used for voltage-controlled oscillators (VCOs) [31], LNAs [32], and other circuits. In this letter, an SISL rectifier based on a high impedance line is proposed at 5.8 GHz. The impedance matching network of the rectifier is uncomplicated and easy to implement. The measured results show that the proposed SISL rectifier achieves an RF-dc conversion efficiency of 3.5% at −30 dBm. Table I shows the estimated theoretical efficiency of an SMS7630 diode at different frequencies calculated by (1) and the comparison between the proposed rectifier and previous work.

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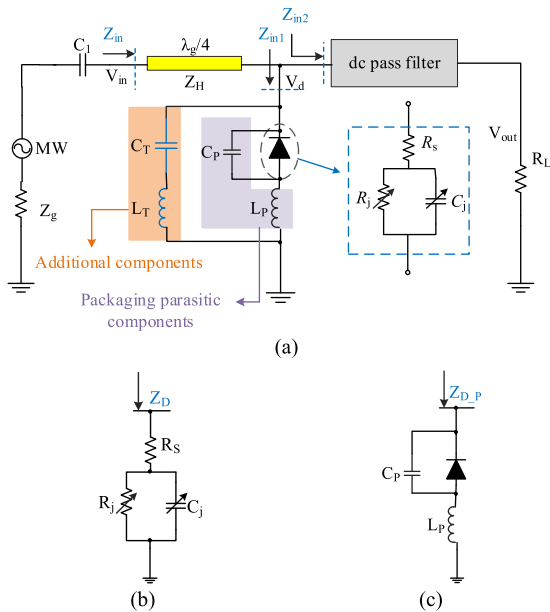


Fig. 1. (a) Schematic of the proposed rectifier. (b) Diode impedance without packaging parameters. (c) Diode impedance with packaging parameters.

TABLE II
SPICE AND PACKAGING PARAMETER OF DIODE SMS7630-079LF

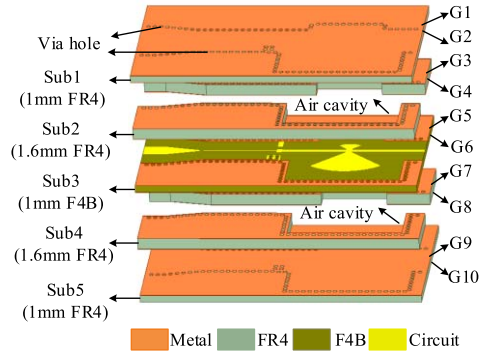


Fig. 2. 3-D view of the proposed rectifier.

Parameter	I_S	R_S	C_{j0}	V_{bi}	L_P	C_P
Unit	A	Ω	pF	V	nH	pF
Value	5×10^{-6}	20	0.14	0.34	0.70	0.16

Fig. 3. Simulated impedance variation with input power in ADS.

II. PRINCIPLE AND DESIGN METHOD

A. Principle

The square-law rectification model presented in [27] can be applied to the efficiency analysis of rectifiers at such a low-power level. Selecting an optimum load and neglecting the matching losses, the estimated theoretical efficiency η can be calculated by [27]

$$\eta = \frac{P_j \cdot \Re_{I0}^2 \cdot R_{j0}}{4} \cdot \frac{1}{\left(1 + (\omega \cdot C_{j0})^2 \cdot R_s \cdot R_{j0}\right)^2} \quad (1)$$

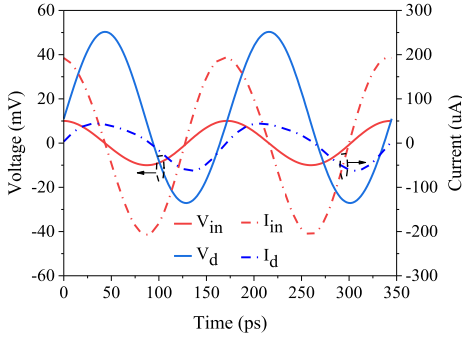


Fig. 5. Simulated voltage and current waveforms of the proposed rectifier at -30 dBm.

from the rectifying diode to the source in a normalized Smith chart. As can be observed in Fig. 5, the peak value of the diode voltage V_d is about four times the input voltage V_{in} at -30 dBm, which contributes to rectifying [34].

B. Rectifier Design

The impedance matching network of the proposed rectifier is specifically designed for the input power level of -30 dBm. At such a low power level, the diode cannot be considered an ideal switch anymore [28]. The Schottky diode impedance Z_D is given by

$$Z_D = R_S + \frac{R_j}{1 + j\omega C_j R_j} \quad (2)$$

where R_j and C_j are the diode resistance and capacitance, respectively. They are approached by their zero-bias values. Considering the parasitic elements L_P and C_P , the whole impedance Z_{D_P} is

$$Z_{D_P} = \frac{Z_D}{1 + j\omega C_P Z_D} + j\omega L_P. \quad (3)$$

After employing the tuning inductor L_T and the dc blocking capacitor C_T , the input impedance Z_{in1} of the rectifying branch is

$$Z_{in1} = \frac{Z_{D_P}(1 - \omega^2 L_T C_T)}{1 - \omega^2 L_T C_T + j\omega C_T Z_{D_P}}. \quad (4)$$

C_T is 3 pF. L_T is 2.09 nH, which is obtained by solving $\text{Im}[Z_{in1}] = 0$. Therefore, Z_{in1} is 755 Ω . The dc-pass filter shows a high impedance at the fundamental frequency. Thus, Z_{in2} is infinity and is neglected.

Then, a quarter-wavelength transmission line is applied to impedance matching. Z_H is the characteristic impedance of the quarter-wavelength transmission line, calculated by

$$Z_H = \sqrt{Z_g Z_{in1}} \quad (5)$$

where Z_g is 50 Ω . Z_H is 194 Ω , which is difficult to implement by a microstrip line.

III. IMPLEMENTATION

The fabricated rectifier consists of five PCBs in Fig. 6. Five PCBs are secured and compacted with screws. The core circuit is located in G5. Substrate F4B is with a loss tangent of 0.002, with a thickness of 1 mm, and a relative dielectric constant of 2.65. It was used as Sub3. Substrate FR4 was used for Sub1, Sub2, Sub4, and Sub5. The thickness is 1 mm for Sub1 and Sub5, and 1.6 mm for Sub2 and Sub4. The fabricated rectifier

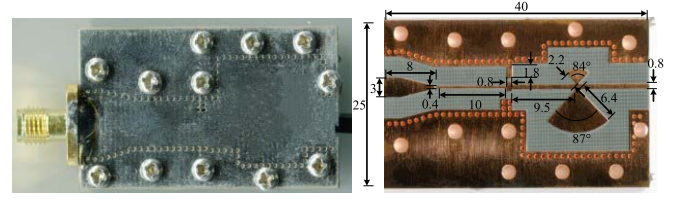


Fig. 6. Photograph of the fabricated rectifier (unit: mm).

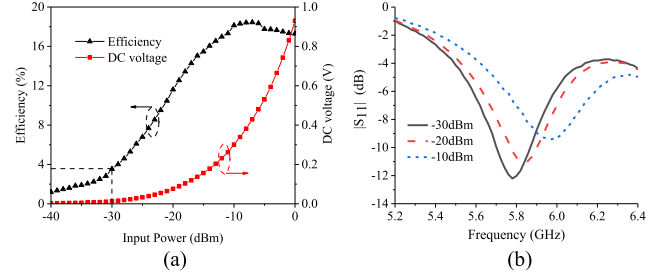


Fig. 7. (a) Measured RF-dc conversion efficiency and output dc voltage versus input power at 5000 Ω load. (b) Measured $|S_{11}|$ versus frequency at different input power level.

is 40 \times 25 \times 6.2 mm. It is a little bulky and complex using the SISL structure compared with microstrip line rectifiers without metal cavity.

A 3 pF capacitor C_1 is used as dc blocking. A short-ended transmission line is applied to realizing inductor L_T . Its width and length are 0.8 and 1.7 mm, respectively. The quarter-wavelength high-impedance transmission line with an optimized width of 0.4 mm serves as the matching network. The dc pass filter is implemented by a quarter-wavelength transmission line and two fan branches. All lengths and widths of microstrip lines were optimized for RF-dc conversion efficiency.

IV. EXPERIMENTAL RESULTS

The proposed rectifier was designed, fabricated, and measured at 5.8 GHz with a 5000 Ω load. A microwave source (Agilent E8267C) was used to generate the RF power. The output dc voltage was measured by a voltage meter. Fig. 7(a) shows the RF-dc conversion efficiency is 3.5%, 11.6%, and 18.1% at the input power at -30 , -20 , and -10 dBm, respectively.

A vector network analyzer (VNA, Agilent N5230A) was applied to measure S-parameter. Fig. 7(b) depicts the measured $|S_{11}|$ from 5.2 to 6.4 GHz at the input power level at -30 , -20 , and -10 dBm, respectively. The optimal frequency is dependent on the input RF power. When the input RF power is -30 dBm, the measured $|S_{11}|$ reaches -12 dB at 5.8 GHz.

V. CONCLUSION

In this letter, a low loss and self-packaged SISL rectifier is proposed. A quarter-wavelength high-impedance transmission line acting as an impedance matching network makes the rectifier facilitated. The proposed topology has a good performance at an input low-power level. The measured efficiency reaches 3.5% at -30 dBm input power. The fabricated rectifier is 40 \times 25 mm. It may be applied to the ultralow power conversion in future low-earth orbit solar power satellite (LEO SPS) demo system.

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